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ERROR PROBABILITIES FOR MAXIMUM LIKELIHOOD DETECTION OF M-ary POISSON PROCESSES IN POISSON NOISE

by

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NATIONAL AERONAUTICS AND SPACE ADMINISTRATION

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SUMMARY

In this problem we have considered some of the recent results in the detection of a Poisson-distributed signal in Poisson noise. Curves for error probabilities are presented for the case of detecting one of M equiprobable signals over a broad range of parameter values. Implicit in these results for system applications is the use of "photon counting" receivers. Attention is given to the optical communication and radar problems for this receiver structure and significant parameters are translated into those used in the report. A complete description of the computational procedures used for making the error probability calculations is given.

INTRODUCTION

The purpose of this paper is to summarize some of the recent results (refs. 1-5) concerning the maximum likelihood detection of Poisson-distributed signals in the presence of Poisson-distributed noise and to tabulate the resultant error probabilities over a broad range of signal and noise when optimum signal design for maximum distance is used. Furthermore, the report shows how these results can be applied to the direct detection of optical signals, with the optimum detector being a counter of photoelectrons. This form of detector can be implemented in the visible portion of the spectrum where photomultipliers exist.

The tabulation is presented in two forms. The first form is related to the detection of M-ary signals and is applicable to the communications problem. The second form is related to the range bin problem in pulsed radar systems. It is felt that the values of the parameters will be applicable to most problems of this type.

BACKGROUND

When a classically describable field is incident upon a photodetecting surface, the probability dp of releasing a photoelectron in an interval dt over a surface area do is (ref 6):

$$dp = \alpha I (t, \sigma) dt d\sigma$$

where I (t,σ) is the intensity of the field and α is equal to η/hf in which η is the quantum efficiency, his Planck's constant, and f is the frequency. For a surface of area A:

$$\alpha$$
 dt $\int_{A} d\sigma I(t,\sigma) = \alpha P(t) dt$

where P (t) is the total collected power at time t. Suppose that the power is related to the incident particle rate n (t) by

$$P(t) = n(t) hf.$$

The probability of releasing a photoelectron in a time dt is then

$$\alpha P$$
 (t) dt = ηn (t) dt.

With this assumption, the probability of releasing K photoelectrons p(K) in a finite interval ΔT is (ref. 7):

$$p(K) = \frac{\left[\int_{t}^{t} + \Delta t \, \eta_{n}(t) \, dt\right]^{K}}{K!} \exp \left[-\int_{t}^{t} \eta_{n}(t) \, dt\right]$$

A system of events obeying this density distribution is called a Poisson process (ref. 8). If the system of events is stationary, we can replace the time average by an ensemble average, as

$$p(K) = \frac{(\eta \overline{n} \Delta T)^{K}}{K!} \exp \left[-\eta \overline{n} \Delta T\right].$$

The system of events would then be called a stationary Poisson process (ref. 8).

Strictly speaking, p(K) is a conditional density and should be written as $p[K/\overline{n}(t)]$ since, in general, n(t) itself is a sample from

a random process. However, in communications, one is interested in "designing" the waveform n(t) to satisfy certain desirable features. Consequently, n(t) is assumed to be deterministic.

Let us assume that we are monitoring the current output of a unity quantum efficiency photodetector in an interval (0, T) and can distinguish all events. Let us further assume that one of two different rates was sent, resulting in one of two received $n_1(t)$, $n_2(t)$, plus a stationary additive constant rate n_1 corresponding to noise. We wish to formulate the maximum likelihood detection procedure for determining which rate is imbedded in the received signal. If the possible transmitted rates are both band-limited to B, then so are $n_1(t)$ and $n_2(t)$. We can therefore partition the (0, T) interval into M subintervals t_{i+1} - t_i = ΔT (ref. 1)

$$0 = t_0 < t_1 < ... < t_{M-1} < t_M = T$$
,

with $\Delta T \leq \frac{I}{2B}$, and consider the quantized version of the possible rates. That is:

$$n_{ji} = \frac{1}{\Delta T} \int_{t_{i}}^{t_{i+1}} n_{j}$$
 (t) dt; $(t_{i} \le t \le t_{i+1}),$ $j = 1, 2.$

To accomplish the detection we consider two hypotheses: H_1 -- rate $n_1(t)$ and noise n_n are present; H_2 -- rate $n_2(t)$ and noise \overline{n}_n are present. We now consider a vector space of M dimensions where each dimension represents the number of events K_1 observed in the corresponding interval. Since the number of events is independent from interval to interval, the vector $K = (K_1, \ldots, K_M)$ has a conditional probability given H_1 of

$$p(\underline{K} \mid H_1) = \prod_{i=1}^{M} \frac{\left[(n_{li} + \overline{n}_{n}) \Delta T \right]^{K_i}}{K_i!} e^{-(n_{li} + \overline{n}_{n}) \Delta T}$$

and a conditional probability, given H2, of

$$p(\underline{K} | H_2) = \prod_{i=1}^{M} \frac{\left[(n_{2i} + \overline{n}_n) \Delta T \right]^{K_i}}{K_i!} e^{-(n_{2i} + \overline{n}_n) \Delta T}$$

The likelihood ratio Λ is then defined as

$$\Lambda(K) = \frac{p(\underline{K}|H_1)}{p(\underline{K}|H_2)} = \prod_{i=1}^{M} \left[\frac{(n_{1i} + \overline{n}_n)}{(n_{2i} + \overline{n}_n)} \right]^{K_i} e^{-(n_{1i} - n_{2i})\Delta T}$$

and the maximum likelihood detection criterion requires, after observing \underline{K} , a comparison of $\Lambda(\underline{K})$ to a threshold c. If $\Lambda \geq c$, we decide rate $n_1(t)$ is imbedded within the received process, and if $\Lambda < c$, we decide rate $n_2(t)$. Since the log function is monotonic, we can also make a decision based upon

$$\log \Lambda \leq \log c$$

where \ denotes the threshold comparison.

Thus $\log \Lambda$ is

$$\sum_{i=1}^{M} K_{i} \log \left[\frac{n_{li} + \overline{n}_{n}}{n_{2i} + \overline{n}_{n}} \right] - \sum_{i=1}^{M} (n_{li} - n_{2i}) \Delta T \geq \log c,$$

which can be rewritten as

$$\Lambda' \leq \left[\log c + \sum_{i=1}^{M} (n_{li} - n_{2i}) \Delta T\right]$$
 (1)

where

$$\Lambda' = \sum_{i=1}^{M} K_{i} \log \left(1 + \frac{n_{1i}}{n_{n}}\right) - \sum_{i=1}^{M} K_{i} \log \left(1 + \frac{n_{2i}}{n_{n}}\right).$$
 (2)

Thus the maximum likelihood detection test involves, equivalently, a comparison of the quantity Λ' in Eq.(2) to the new threshold in Eq.(1).

"Distance" Considerations

Let us first consider the case where $n_{2i} = 0$ for all i and we are detecting the rate $n_i(t) + n_n$ versus \overline{n}_n alone (i.e., signal to no signal). Then the test in Eq. (1) becomes

$$\Lambda' = \sum_{i=1}^{M} K_{i} \log \left(1 + \frac{n_{1i}}{n_{n}}\right) \ge \log c + \sum_{i=1}^{M} n_{1i} \Delta T$$

Clearly, since K_i is Poisson-distributed:

$$E_{\underline{K}} \left[\Lambda' / H_{1} \right] = \sum_{i=1}^{M} (n_{1i} + \overline{n}_{n}) \log \left(1 + \frac{n_{1i}}{\overline{n}_{n}}\right) \Delta T$$

and

$$E_{\underline{K}}[\Lambda'/H_2] = \sum_{i=1}^{M} \frac{1}{n_i} \log \left(1 + \frac{n_{1i}}{n_n}\right) \Delta T.$$

We define the "distance" D between the two hypotheses as

$$D = E_{\underline{K}} [\Lambda'/H_1] - E_{\underline{K}} [\Lambda'/H_2]$$
$$= \sum_{i=1}^{M} n_{1i} \log (1 + \frac{n_{1i}}{\overline{n}_n}) \Delta T.$$

We would like to investigate maximum and minimum of the distance

$$D = \sum_{i=1}^{M} n_{li} \log \left[1 + \frac{n_{li}}{\overline{n}_{n}} \right] \Delta T$$

subject to the energy constraint

$$\sum_{i=1}^{M} n_{li} = K' = \frac{K_s}{\Delta T} .$$

Using a Lagrange multiplier we seek the minimum of

$$I = \Delta T \sum_{i=1}^{M} n_{li} \log \left[1 + \frac{n_{li}}{\overline{n}_{n}}\right] + \lambda \sum_{i=1}^{M} n_{li}$$

which requires

$$\frac{\partial I}{\partial n_{li}} = \Delta T \log \left[1 + \frac{n_{li}}{\overline{n}_{n}}\right] + \frac{\Delta T n_{li}}{n_{li} + \overline{n}_{n}} + \lambda = 0$$

and is a set of M equations that must be satisfied for all i. The solution is $n_{1i} = n$ for all i yielding

$$D_{\min} = \sum_{i=1}^{M} n_{li} \log \left[1 + \frac{n_{li}}{n_{n}}\right] \Delta T$$

$$= \Delta T \operatorname{Mn} \log \left[1 + \frac{n}{\overline{n}}\right] = K_{s} \log \left[1 + \frac{K_{s}}{T\overline{n}}\right]$$

which can be identified as a minimum.

The maximum value can be obtained by noting that

$$\sum_{i=1}^{M} n_{li} \log \left[1 + \frac{n_{li}}{\overline{n}_{n}}\right] \leq \left[\sum_{i=1}^{M} (n_{li})^{2}\right]^{1/2} \left[\sum_{i=1}^{M} \log^{2} \left(1 + \frac{n_{li}}{\overline{n}_{n}}\right)\right]^{1/2}$$

$$\leq \left[\sum_{i=1}^{M} n_{li}\right] \left[\sum_{i=1}^{M} \log\left[1 + \frac{n_{li}}{n_{n}}\right]\right]$$

with the equality occurring for $n_{ij} = K' \delta_{ij}$.

The maximum value for the sum becomes:

$$D_{\max} = \sum_{i=1}^{M} n_{li} \log \left[1 + \frac{n_{li}}{n_{n}}\right] \Delta T = K_{s} \log \left[1 + \frac{K_{s}}{n_{n}\Delta T}\right]$$

Thus, the distance D is maximized when the rate $n_1(t)$ is concentrated in one ΔT interval. Hence, the signal maximizing detectability* (refs. 1,2) also maximizes distance. Notice also, that for the important case when $\overline{n}_n >> n_1$, for all i,

$$D = \frac{\sum_{i=1}^{M} n_{i}^{2} \Delta T}{\overline{n}_{n}}$$

with distance and detectability identical.

The maximization can also be obtained for the case $n_{2i} \neq 0$.

$$D = E_{\Lambda} \lceil \Lambda' / H_1 \rceil - E_{\Lambda'} \lceil \Lambda' / H_2 \rceil = \sum_{i=1}^{M} (n_{1i} - n_{2i}) \log \left(\frac{n_{1i} + \overline{n}_n}{n_{2i} + \overline{n}_n} \right) \Delta T$$

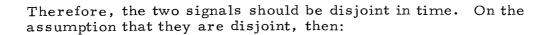
$$= \sum_{i=1}^{M} \log \left[\frac{n_{1i} + \overline{n}_{n}}{n_{2i} + \overline{n}_{n}} \right]^{(n_{1i} - n_{2i})} \Delta T.$$

Since n_{li} and n_{2i} must be non-negative, the log term is maximized for all i by having $n_{2i} = 0$ when n_{li} is not, and vice versa. This maximizes

$$\left[\begin{array}{c} \frac{n_{1i} + \overline{n}_{n}}{n_{2i} + \overline{n}_{n}} \end{array}\right]^{(n_{1i} - n_{2i})}$$

by giving the largest magnitude to both the bracketed term and the exponent simultaneously and leaves the sign of the log positive.

^{*} Signal-to-noise ratio



$$D = \sum_{i=1}^{M} n_{1i} \log \left[1 + \frac{n_{1i}}{\overline{n}_{n}}\right] + \sum_{i=1}^{M} n_{2i} \log \left[1 + \frac{n_{2i}}{\overline{n}_{n}}\right].$$

To maximize D, we have only to notice that this is identical to maximizing each signal independently or to concentrate each signal in a different time interval. The optimum processor, therefore, calculates

$$\Lambda' = \sum_{i=1}^{M} K_i \log \left[1 + \frac{n_{li}}{\overline{n}_n}\right] - \sum_{i=1}^{M} K_i \log \left[1 + \frac{n_{2i}}{\overline{n}_n}\right]$$

which is compared to the threshold

$$\log c + \sum_{i=1}^{M} (n_{1i} - n_{2i}) \Delta T.$$

If the two signals are equiprobable with equal energy

$$\begin{array}{ccc}
M & & M \\
\sum_{i=1}^{N} n_{1i} \Delta T = \sum_{i=1}^{N} n_{2i} \Delta T,
\end{array}$$

c=land the threshold is zero. The H_1 , H_2 log likelihood functions are calculated for each choice of waveform n_i (t) and the hypothesis is selected according to the largest result.

If M waveforms are used, the optimum processor would calculate the likelihood function for each of the M choices and select the maximum. Expressed mathematically, one obtains for optimum signal design under an equiprobable, equal energy assumption

$$\Lambda' = K_{i} \log \left[1 + \frac{K_{s}}{\overline{n}_{n} \Delta T}\right] - K_{j} \log \left[1 + \frac{K_{s}}{\overline{n}_{n} \Delta T}\right] \leq 0$$

or
$$K_i \geq K_j$$
.

Hence, the problem is reduced to counting the number of photoelectrons in each ΔT interval and selecting the interval with the largest count. The probability of correct detection P_D is then:

 $P_D = [Probability that K_j > K_i for all i/n_j is the transmitted waveform].$

$$+ \sum_{r=2}^{M} \frac{1}{r} \left[\text{Probability that } K_{j} = K_{i} \text{ for } r-1 \text{ intervals/} n_{j} \text{ is the transmitted waveform} \right].$$

This is then averaged over all choices of waveform, which for the equiprobable set is just $\,P_D^{}$.

This can be written

$$P_{D} = \sum_{x=1}^{\infty} \left\{ \frac{\left(K_{s} + \overline{n}_{n} \Delta T\right)^{x}}{(x)!} e^{-\left(K_{s} + \overline{n}_{n} \Delta T\right)} \left[\sum_{i=0}^{x-1} \frac{\left(\overline{n}_{n} \Delta T\right)^{i}}{i!} e^{\left(\overline{n}_{n} \Delta T\right)}\right]^{M-1}$$

$$\left[\frac{\left(1+B\right)^{M}-1}{MB}\right] + \frac{1}{M} e^{-\left(K_{s} + \overline{n}_{n} M \Delta T\right)} \tag{3}$$

where

$$B = \frac{\left(\overline{n}_{n} \Delta T\right)^{x}}{x \cdot 1} \frac{\left(\overline{n}_{n} \Delta T\right)^{i}}{\left(\overline{n}_{n} \Delta T\right)^{i}}$$

$$i = 0$$

$$i \cdot !$$

The error probability $P_E = 1 - P_D$.

Presentation of the Data

The data are presented in two forms. In the first form (Program 1, Figures 1-10) P_E is plotted versus M for various values of K_S and $K_n=\overline{n}_n$ ΔT . These parameters are directly related to the received signal and noise energies, respectively. Thus, if one determines from the range equation that P_S is the received signal peak power, then

$$K_s = \frac{\eta P_s \Delta T}{hf}$$

whereas if the received average noise power is Pn,

$$K_{n} = \frac{\eta P_{n} \Delta T}{hf} \qquad (4)$$

If non-diffraction-limited collecting optics is used, P_n is calculated as

$$P_{N} = A_{R} T_{O} \Omega_{r} N_{\lambda} \Delta \lambda \tag{5}$$

with

 N_{λ} = spectral radiance (power/unit area, solid-angle bandwidth)

T_o = fraction of optical transmission through all elements

 A_R = area of collector

 Ω_r = resolution of receiving collector (solid angle)

 $\Delta \lambda = \text{optical bandwidth.}$

In the visible region of the spectrum 5-6000 Å, $\rm K_n$ for non-diffraction-limited optics with a blue sky background can be written as

$$K_n = 8\eta\Delta T(\alpha D)^2 T_o(\Delta \lambda)$$

_PROGRAM I

	HÜRW REQ	04/30/68 1,C76501,T5 UEST TAPE7,	00,CM50000 LO. PLO	80/80 NASA T SAO 568	101074	KARP3	HURWITZ	
	RUN (5,,,,,,7777	77)		- .			
	LGO.	DOUBLE PR	ARP3 (INPU	B, EKSN, EL	.10,ELM,	EM,EXK,F	D,PE,PI,PX	,
	÷	2, XK	SAVE,SUM,SI ,XKN,XKNMA OUT,IPLT,TI	X,XKNMIN,	XK5 • XKS	IMX•MX•N3		
		DATA IXMA	YAX/8.,11. X,XKNMAX,XI		MIN/100	00,3.D+2,	1.D-30,1.D	-12/
	CCC.		IITZ UE OF DIFF! ED IN WRITI		TWEEN T	WO CONSE	CUTIVE TER	MS -
	222	"EKSN=EXP(-XKSN) TANT USED I		NG MAGN	 LITUDES C	F NUMBERIC	AL VALUES
	CCC	ĒM=INITIA	L VALUE OF FOR LOG(M)		I KS ANT) EACH K	Ma.x	
	CCC	EXK=INPUT	OF K					
	222 222 222 222	IEND=TEST IEND MUST	ED TO INSUIT VALUE FOR BE NINE FOR	STOPPIÑO OR ALL BU	RUN - IT LAST	•		CULATION
,	CCC.		NUMBER_TO			PE '_S		
	222 222 222	JMP MUST JMP=TEST	BE ZERO FOI VALUE FÖR I OF VALUES	READING V				
	CCC		OF VALUES		•			CONTRACTOR OF THE PARTY OF THE
	ccc ccc	N1 = NN + 1	UNTS NUMBEI	R OF M'S	FOR EAC	TH K		
	ccc ccc	N2=NN+2 PD DESCRI PE=1-PD	BED IN WRI	r <u>e up</u>				
	CCC)*EKSN/(I KSN	FACTORIAL	.)			
	CCC CCC	SAVE LAST) TEST MAGN VALUE OF TION (KN**	rerm				
• •	ccc		MATION (PI			-		
	CCC	TERM2=SUM	ERM2*TERM3 PI**(M-1)					
	CCC	TOTAL=SUM	B)**M-1/(M: MATION(TERM FOR LOG(PE	۷)	H KS AN	ID FACH K	,	
	ccc ccc		OR LOG(M) F					
=	CCC CCC		X VALUE OF N VALUE OF					
	ccc	XKS=KS	,,		-	•		
_								

0	4/30/68 80/80 LIST
ccc	XKSN=KS+KN
ccc	XM=EM**I
CCC	XMB=XM*B
ccc	XMBMIN≃TEST VALUE FOR XMB
CCC	XM1=XM-1
ccc	XX=IX=INDEX OF OUTER LOOP
CCC	Y=ARRAY FOR LOG(PE) FOR ALL CURVES FOR EACH KS
	DIMENSION X(3000), Y(3000), EX(100), WY(100), NPT(100), EXK(100)
	EL10=300.D0*DLOG(10.D0)
	CALL INITPLT(IPLT)
1	READ (IN, 3) EM, XKS, NM, NK, IEND, JMP
	L=0 IF (JMP •NE•0) GO TO 2
	READ (IN,4) (EXK(I), I=1,NK) FORMAT (D10.3)
ccc ¯	LOOP FOR K
	DO 70 J=1,NK
_	NPT(J)=0
3	FORMAT (2D10 • 3 • 415)
	XKN=EXK(J)
CCC	LOOP FOR M
10	DO 65 I=1,NM
CCC	PRINT HEADINGS FIRST TIME THROUGH THIS LOOP
	IF (I.NE.1) GO TO 14
	WRITE(NOUT, 12) XKS, XKN
12	FORMAT (1H1,40X,3HKS=,D9,2,10X,3HKN=,D9,2//)
	WRITE (NOUT,13)
	FORMAT (17X,1HM,20X,2HPD,20X,2HPE,20X,6HLOG(M),16X,7HLOG(PE)//)
14	XM=EM**I
CCC	XM1=XM-1.DO
CCC	TEST TO INSURE EXP(KN) DOES NOT EXCEED MACHINE CAPACITY IF KN GT XKNMAX, INCREASE M AND CONTINUE
	IF (XKN.GT.XKNMAX) GO TO 65
	XKSN=XKS+XKN
CCC	TEST VALUE TO INSURE NO PREMATURE CUTOFF
	ICUTOF=2.DO*XKSN
	EKSN=DEXP(-XKSN)
CCC	FORMATION OF PX FOR X=1
	PX=XKSN*EKSN
	SUM=1.DO
	T=1.D0
CCC	FORMATION OF B FOR X=1
	B=XKN
	XMB=XM*B
CCC	TEST FOR (1+B)**M TOO LARGE
	S=DLOG(1.D0+3)
	S=EL10/XM-S
	IF (\$.6T.0.D0) GO TO 17
ccc	VALUE FOR (1+B)**M TOO LARGE FOR MACHINE
	TERM3=0.D0
ccc	GO TO 16
CCC	VALUE FOR (1+B)**M SUFFIENTLY SMALL
ccc	TERM3=((1.00+B)**XM-1.00)/XMB TEST FOR KN NEAR ZERO
	IF (XKN.GE.XKNMIN) GO TO 16
1.5	TERM2=1.00
•/-	TERM3=1.D0

0	4/30/68 - 80/80 LIST	
. 0	GO TO 20	
ccc	FORMATION OF PI, SUMPI, TERM2 FOR X=1	
16	PI=DEXP(-XKN)	
	SUMPI=PI TERM2=PI**XM1	
ccc	LOOP FOR SUMMATION OF TERM	
20	TERM=PX*TERM2*TERM3	
	TOTAL=TERM	
	DO 50 IX=2,IXMAX	
	SAVE=TERM XX=IX	
	X1=XX-1.D0	
ccc	FORMATION OF PX FOR X.GE.2	
	PX=PX*XKSN/XX	
ccc	TEST FOR KN NEAR ZERO	
ccc	IF (XKN.LT.XKNMIN) GO TO 45	
ccc	FORMATION OF PI FOR X.GE.2 PI=PI*XKN/X1	
ccc	FORMATION OF TERM2 FOR X.GE.2	
ccc	FORMATION OF B FOR X.GE.2	
-	SUMPI=SUMPI+PI	
25	TERM2=(SUMPI)**XM1	
	T=T*XKN/X1	
	A=SUM SUM=SUM+T	
	B=XKN*B*A/(XX*SUM)	
CCC	TEST FOR (1+B)**M TOO LARGE	
	S=DLOG(1.D0+B)	
	S=EL10/XM-S	
c c c	IF (S.LT.0.D0) GO TO 50	
CCC	FORMATION OF TERM3 TEST FOR XMB LE XMBMIN	
CCC	IF MB LE XMBMIN APPROX TERM3 WITH FIRST TWO TERMS OF	MŘ ONLY
35	XMB=XM*B	TIB ONE!
	IF (XMB.GT.XMBMIN) GO TO 40	
	TERM3= $1 \cdot D0 + (XM-1 \cdot D0) *B/2 \cdot D0$	
	GO TO 45	
CCC	TERM3=((1.D0+B)**XM-1.D0)/XMB FORMATION OF TERM FOR X.GE.2	
	TERM=PX*TERM2*TERM3	
ccc	SUMMATION OF TERMS	
	TOTAL=TOTAL+TERM	
CCC	TEST IF DIFFERENCE OF TERMS FOR X=N AND X=N+1 IS SUF	SMALL
	A=DABS(TERM-SAVE)	
ccc	IF (A.LT.TEST.AND.IX.GT.ICUTOF) GO TO 55 TEST TO INSURE AGAINST PREMATURE CUTOFF	
	CONTINUE	
ccc	COMPUTATION OF FINAL VALUE FOR A GIVEN M	
55	PD=TOTAL+DEXP(-(XKS+XM*XKN))/XM	-
	PE=1 •D0-PD	
	L=L+1	
	ELM=DLOG10(XM) X(L)=SNGL(ELM)	
	ELM=DLOG10(PE)	
	Y(L)=SNGL(ELM)	
	NPT(J)=NPT(J)+1	
	WRITE (NOUT, 60) XM, PD, PE, X(L), Y(L)	·-

04/30/68 80/80 LIST	
60 FORMAT (9X,3(D16,9,6X),2(F16,9,6X))	
65 CONTINUE	
70 CONTINUE	
L1=L+1	
L2=L+2	
X(L1)=0.	
X(L2)=0•	
Y(L1)=0.	
Y(L2)=0•	
CCC PLOTTING ROUTINES FOR CAL-COMP PLOTTER	
CALL PLOT(0.,-31.,-3)	
CALL PLOT(0.,2.,-3)	
CALL SCALE (X,XAX,L,1)	
CALL SCALE (Y,YAX,1,1)	
CALL AXIS (0.,0.,6HLOG(M),-6,XAX,0.,X(L1),X(L2))	
CALL AXIS (0.,0,7HLOG(PE),7,YAX,90.,Y(L1),Y(L2))	
I1=0	
DO 90 I=1,NK	
NN=NPT(I)	
DO 85 J=1,NN	
11=11+1	
EX(J)=X(I1)	
85 WY(J)=Y(I1)	
N1=NN+1	
N2=NN+2	
EX(N1) = X(L1)	
$= \frac{EX(N2) + X(L2)}{EX(N2) + X(L2)}$	
WY(N1) = Y(L1)	
WY(N2)=Y(L2)	-
CALL LINE (EX, WY, NN, 1, 0, 0)	
90 CONTINUE	
CALL SYMBOL(0.5,16.0,.3,3HKS=,0.,3)	
CALL NUMBER(1.5,16.0,.3,XKS,0.,-1)	_
CALL PLOT(20.,0.,-3)	
110 CONTINUE	_
IF(IEND.EQ.9) GO TO 1	
CALL FIN(IPLT)	_
120 STOP	
END	

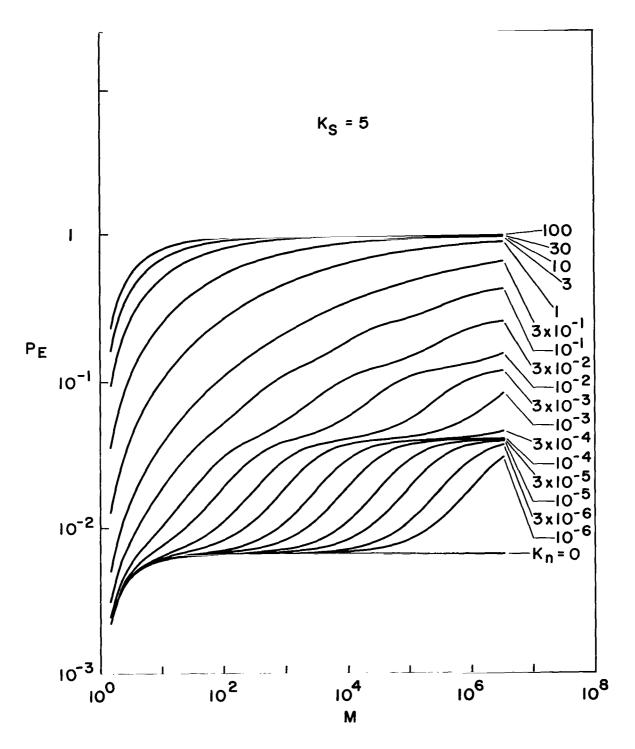


Figure 1.- Error probabilities for $\overline{n}_n \Delta T = K_n$ fixed as a function of M

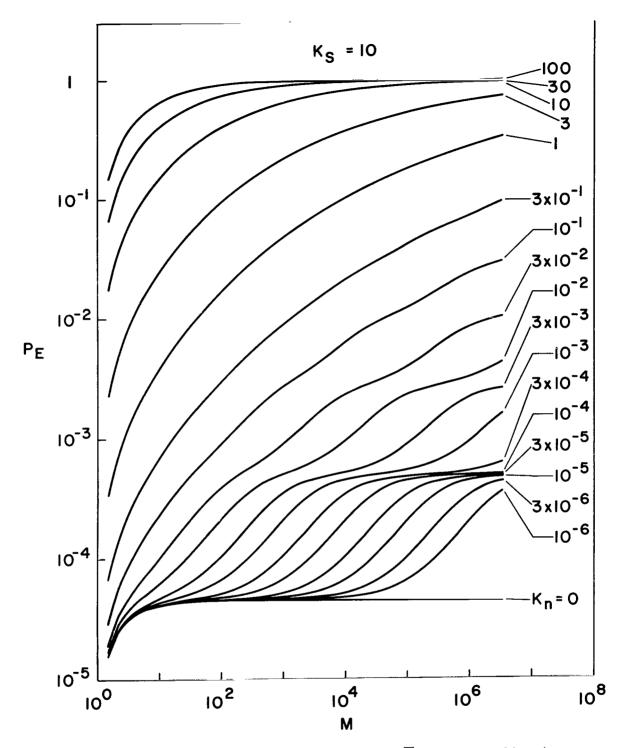


Figure 2.- Error probabilities for $\overline{n}_n \Delta T = K_n$ fixed as a function of M

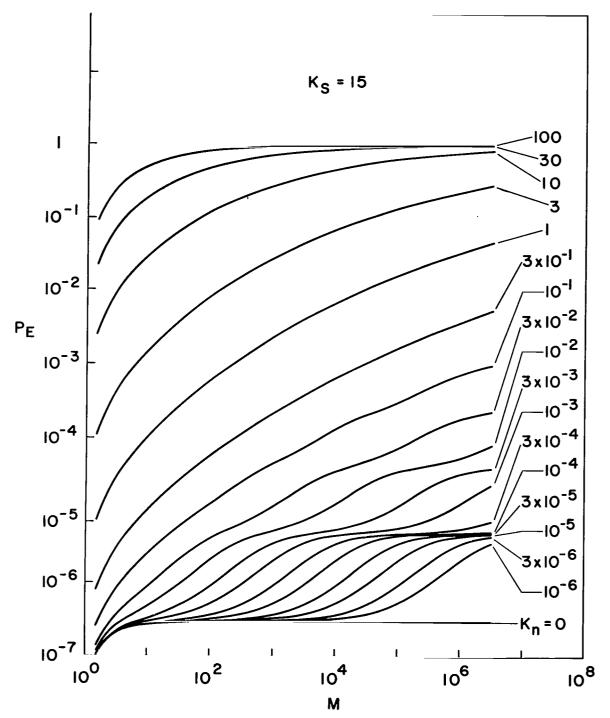


Figure 3.- Error probabilities for $\overline{n}_n \Delta T = K_n$ fixed as a function of M

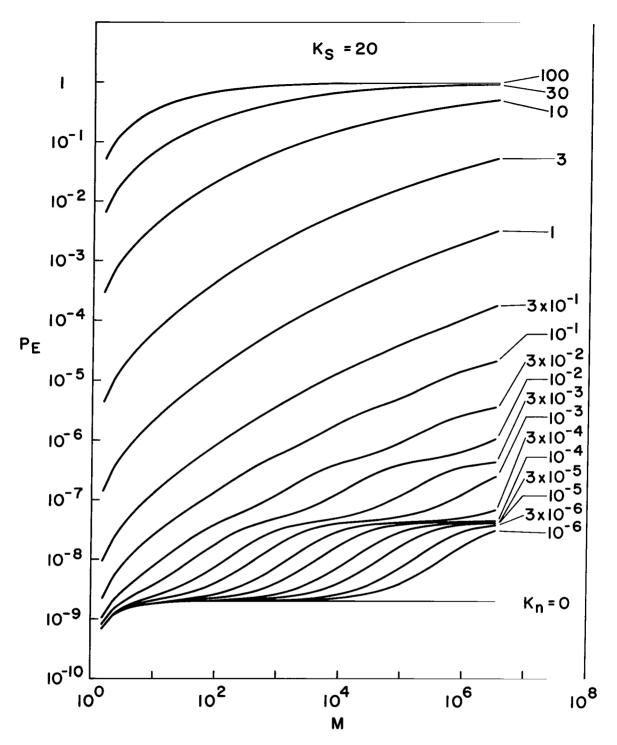


Figure 4.- Error probabilities for $\overline{n}_n \Delta T = K_n$ fixed as a function of M

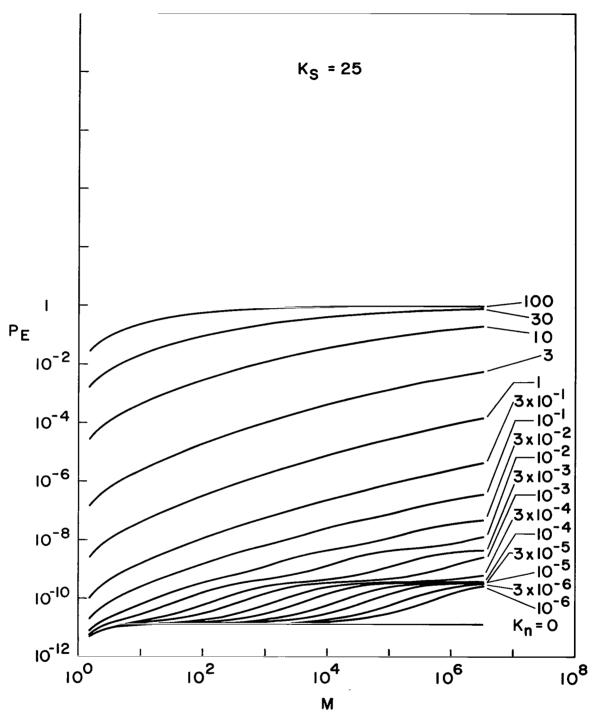


Figure 5.- Error probabilities for $\overline{n}_n \Delta T = K_n$ fixed as a function of M

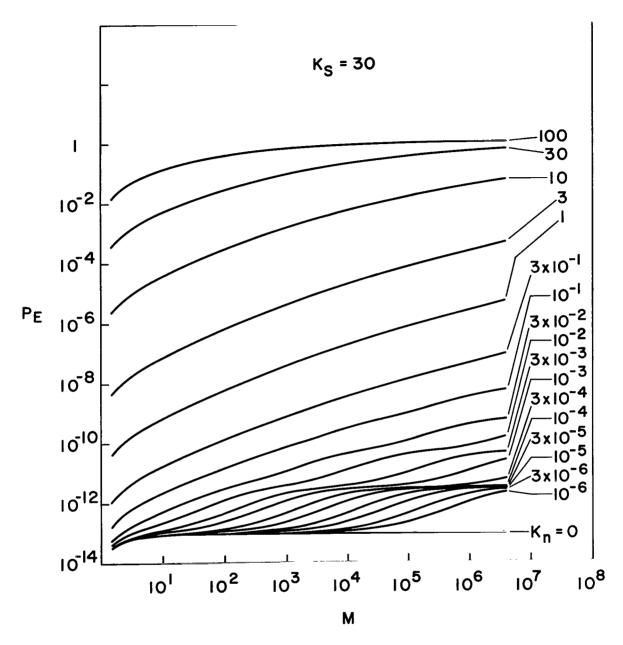


Figure 6.- Error probabilities for $\overline{n}_n \Delta T = K_n$ fixed as a function of M

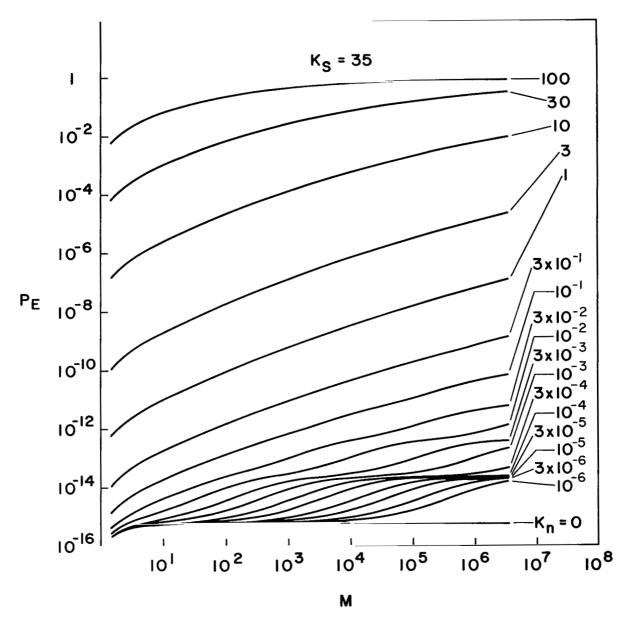


Figure 7.- Error probabilities for $\overline{n}_n \Delta T = K_n$ fixed as a function of M

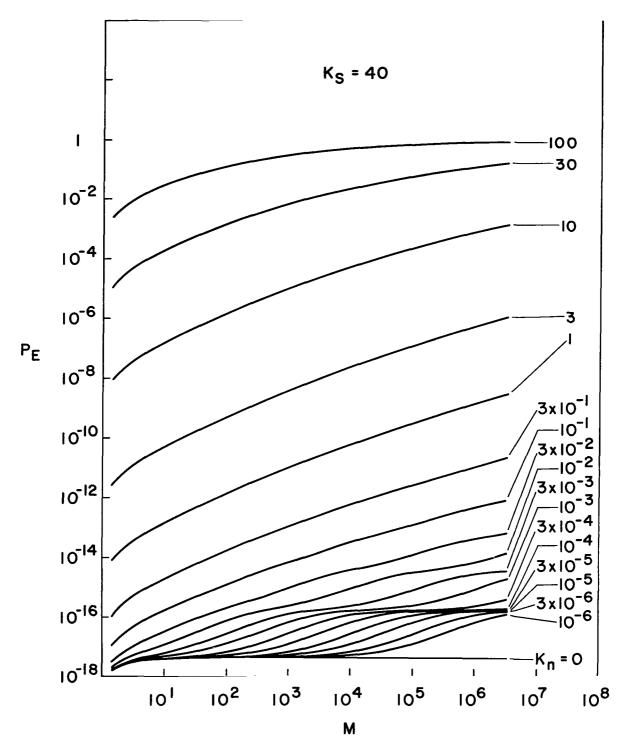


Figure 8.- Error probabilities for $\overline{n}_n \Delta T = K_n$ fixed as a function of M

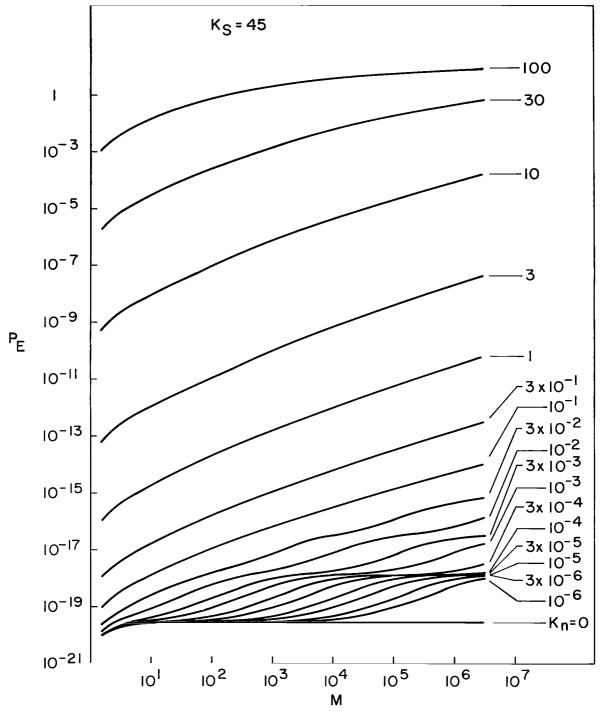


Figure 9.- Error probabilities for $\overline{n}_n \Delta T = K_n$ fixed as a function of M

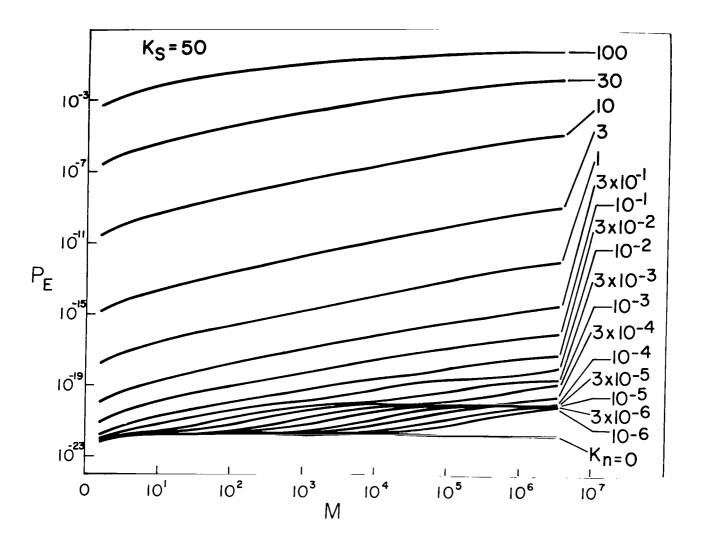


Figure 10.- Error probabilities for $\overline{n}_n \Delta T = K_n$ fixed as a function of M

where $\Delta\lambda$ is in angstroms, α is the resolution in arc seconds, and D is the diameter of the collector in centimeters. For diffraction-limited optics, α is related to D by

$$\alpha = 2.44 \times 10^5 \frac{\lambda}{D}$$
 arc seconds;

hence

$$\mathrm{K_n} = 4.76\,\eta\,\Delta\mathrm{T\,T_o}(\Delta\lambda)\lambda^2 \times 10^{11}$$

with λ the wavelength in centimeters.

In the second form (Program 2, Figures 11-20) we consider the sampling interval T to be a constant and examine $P_{\overline{L}}$ as M becomes large, or since $\Delta T = T/M$, as ΔT becomes small. The assumption is that the signal energy can always be concentrated in the ΔT interval. For this case we use the parameters K_{S} and $K=\overline{n}_{n}T$. The noise in the ΔT interval is then K/M. K_{S} can be calculated as before while K is calculated as

$$K = \eta \frac{P_n T}{hf} .$$

It can be shown analytically that as $\,M\to\infty$, an asymptotic value for $\,P_E\,$ is reached, where (ref. 5)

$$\lim_{M \to \infty} P_{E} = \left[1 + K_{s} \left[\frac{K - 1 + e^{-K}}{K} \right] \right] e^{-K_{s}}$$

This asymptote is quite apparent in Figures 11-20 and is seen to vary with K as expected. Physically, this implies that if bandwidth and computation are expedient, one can always approach

$$P_{E} = (1 + K_{s}) e^{-K_{s}}$$

independent of the noise background by using narrower intervals. This procedure also provides better range resolution for the radar case. Implicit in these calculations is the fact that \overline{n}_n is a constant or that the optical filter bandwidth is generally quite large.

PROGRAM 2

		80/80 LIST 2,C76501,T500,CM50000. NASA I01074 KARP4 HURWITZ
		JEST TAPE7,LO. PLOT SAO 766 RING IN.
	RE613	(0000 • , 77777)
		PROGRAM KARP4 (INPUT,OUTPUT,TAPE7,TAPE5=INPUT,TAPE6=OUTPUT)
		DIMENSION EX(100), EXK(100), NPT(100), WY(100), X(3000), Y(3000)
		DOUBLE PRECISION A,B,EKSN,EL10,ELM,EM,EXK,PD,PE,PI,PX,
		1 S,SAVE,SUM,SUMPI,T,TERM,TERM2,TERM3,TEST,TOTAL,X1,
		Ž XK, XKN, XKNMAX, XKNMIN, XKS, XKSN, XM, XM1, XMB, XMBMIN, XX
		DATA IN, NOUT, IPLT, TEST/5, 6, 7, 1. D-24/
		DĀTĀ XAX,YAX/11.514./
		DATA IXMAX,XKNMAX,XKNMIN,XMBMIN/1000,3.D+2,1.D-30,1.D-12/
	CCC	KARP/HURWITZ
	CCC	A=ABS.VALUE OF DIFFERENCE BETWEEN TWO CONSECUTIVE TERMS
	CCC	B DESCRIBED IN WRITE UP
	CCC CCC	EKSN=EXP(-XKSN) ELIO-CONSTANT USED FOR TESTING MAGNITUDES OF NUMBERICAL VALUES
	CCC	ELM=LOG(M)
	-ccc -	EM=LOOTM)
	CCC	EXK=INPUT OF K
•	CCC	EX=ARRAY FOR LOG(M) FOR EACH KS AND EACH K
	CCC	I = INDE X
	ĊCC	II COUNTS NUMBER TOTAL NUMBERS OF PE'S
	CCC	ICUTOF USED TO INSURE NO PREMATURE CUT-OFF OF THE CALCULATION
	CCC	TEND=TEST VALUE FOR STOPPING RUN
	. C C C	IEND MUST BE NINE FOR ALL BUT LAST DATA CARD IXMAX=MAX VALUE FOR INDEX IX
	CCC	J=INDEX
	CCC	JMP=TEST VALUE FOR READING VALUES OF K
	CCC	JMP MUST BE ZERO FOR FIRST DATA CARD ÑK≔NUMBER OF VALUES OF K
	~CCC	NM=NUMBER OF VALUES OF M
	CCC	NN=NPT(I)
	CCC	NPT(J) COUNTS NUMBER OF M'S FOR EACH K
	. ccc .	N1=NN+1
	ccc	N2=NN+2
	-CCC	PD DESCRIBED IN WRITE UP
	CCC	PE=1-PD
	CCC	PI=(KN**I)*EKSN/(I FACTORIAL)
	CCC	PX=XKSN*EKSN
	CCC	S USED TO TEST MAGNITUDES
	CCC	SAVE LAST VALUE OF TERM
	CCC	SUM=SUMMATION (KN**I/I)
	CCC	SUMPI=SUMMATION (PI)
	CCC	T=DUMMY
	CCC	TERM=PX*TERM2*TERM3
	CCC	TERM2=SUMPI**(M-1) TERM3=(1+B)**M-1/(M*B)
	CCC	TOTAL=SUMMATION(TERM)
	CCC	WY=ARRAY FOR LOG(PE) FOR EACH KS AND EACH K
	ČCC	X=ARRAY FOR LOG(M) FOR ALL CURVES FOR EACH KS
	CCC	XK=EXK(J)
	CCC	XKN=KN
	ccc	XKNMAX = MAX VALUE OF KN
	ĈĊĊ Ċ	XKNMIN=MIN VALUE OF KN

		0.00 - 107		
	ĊCC 04	4/30/68 _ 80/80 LIST XKS=KS		
	ccc	XKSN=KS+KN		
	CCC	XM=EM**I		
	CCC	XMB=XM*B		
	CCC	XMBMIN=TEST VALUE FOR XMB XM1=XM-1		
	CCC	XX=IX=INDEX OF OUTER LOOP		
	ccc	Y=ARRAY FOR LOG(PE) FOR ALL CURVES FOR EACH KS		
		EL10=300.D0*DLOG(10.D0)		
		CALL INITPLT(IPLT)		
	1	READ (IN,3) EM,XKS, NM,NK, TEND, JMP		
		L=0 IF (JMP •NE•0) GO TO 2	-	
		READ $(1N,4)$ $(EXK(1), I=1,NK)$		
	4	FORMAT (D10.3)		
	ccc	LOOP FOR K		
	2	DO 70 J=1,NK		
_		NPT(J)=0		
	3	FORMAT (2D10.3,4T5)		
	CCC	XK=EXK(J) LÕOP FOR M		
		DO 65 I=1,NM		
		PRINT HEADINGS FIRST TIME THROUGH THIS LOOP		
		IF (I.NE.1) GO TO 14		
		TWRITE (NOUT, 12) XKS,XK	•	
	12	FORMAT(1H1,40X,3HKS=,D9.2,10X,2HK=,D9.2//)		
	• 0	WRITE (NOUT, 13)	DE1/	<i>,</i> ,
		FORMAT(17x,1HM,20x,2HPD,20x,2HPE,20x,6HLOG(M),16x,7HLOG(XM=EM**I	- L) / /	, -
	14	XM1=XM-1.D0		
-		XKN=XK7XM		
	CCC	TEST TO INSURE EXP(KN) DOES NOT EXCEED MACHINE CAPACITY		
	CCC	IF KN GT XKNMAX, INCREASE M AND CONTINUE		
_		IF (XKN.GT.XKNMAX) GO TO 65		
	666	XKSN=XKS+XKN TEST VALUE TO INSURE NO PREMATURE CUTOFF		
	ccc -	ICUTOF=2.DO*XKSN		
		EKSN=DEXP(-XKSN)		
	CCC .	FORMATION OF PX FOR X=1		
		PX=XKSN*EKSN		
		ŜUM=1•D0		
		T=1.00		
	ccc	FORMATION OF B FOR X=1 B=XKN		
-	-	XMB=XM*B		
	ccc	TEST FOR (1+B)**M TOO LARGE		
		S=DLOG(1.D0+B)		
		S=EL10/XM-S		
	•	IF (S.GT.O.DO) GO TO 17		
	CCC	VALUE FOR (1+B) **M TOO LARGE FOR MACHINE		
		TERM3=0•D0 GO TO 16		
	CCC	VALUE FOR (1+B) **M SUFFIENTLY SMALL		
		TERM3=((1.D0+B)**XM-1.D0)/XMB		
-	CCC	TEST FOR KN NEAR ZERO		
_	15	IF (XKN.GE.XKNMIN) GO TO 16		
		TERM2=1.DO		

0	4/30/68 80/80 LIST
	TERM3=1 • DO
	GO TO 20
CCC	FORMATION OF PI,SUMPI,TERM2 FOR X=1
	PI=DEXP(-XKN)
	SUMPI=PI
	TERM2=PI **XM1
	LOOP FOR SUMMATION OF TERM
20	TERM=PX*TERM2*TERM3
	TOTAL=TERM
	DO 50 IX=2,IXMAX
	SAVE=TERM
	X X = I X
	X1=XX-1.D0
ccc	FORMATION OF PX FOR X.GE.2
	PX=PX*XKSN/XX
	TEST FOR KN NEAR ZERO
ccc	
	ÎF (XKN-LT-XKNMIN) GO TO 45
CCC	FORMATION OF PI FOR X.GE.2
	PI=PI*XKN/XI
CCC	FORMATION OF TERM2 FOR X.GE.2
CCC	FORMATION OF B FOR X.GE.2
	SUMPI=SUMPI+PI
25	TERM2=(SUMPI)**XM1
_	T=T*XKN/X1
	A=SUM
	SUM=SUM+T
	B=XKN*B*A/(XX*SUM)
555	TEST FOR (1+B)**M TOO LARGE
CCC	
	S=DLOG(1.DO+B)
	S=EL10/XM-S
	IF (5.LT.0.DO) GO TO 50
CCC	FORMATION OF TERM3
CCC	TEST FOR XMB LE XMBMIN
CCC	IF MB LE XMBMIN APPROX TERM3 WITH FIRST TWO TERMS OF MB ONLY
35	XMB=XM*B
-	IF (XMB.GT.XMBMIN) GO TO 40
	TERM3= 1.D0+(XM-1.D0)*B/2.D0
	GO TO 45
	TERM3=((1.DO+B)**XM-1.DO)/XMB
CCC; _	FORMATION OF TERM FOR X.GE.2
	TERMEPX*TERM2*TERM3
CCC	SUMMATION OF TERMS
	TOTAL-TOTAL-TERM
CCC	TEST IF DIFFERENCE OF TERMS FOR X=N AND X=N+1 IS SUFF. SMALL
	A=DABS(TERM-SAVE)
	IF (A.LT.TEST.AND.IX.GT.ICUTOF) GO TO 55
CCC	TEST TO INSURE AGAINST PREMATURE CUTOFF
	CONTINUE
	COMPUTATION OF FINAL VALUE FOR A GIVEN M
	PD=TOTAL+DEXP(-(XKS+XM*XKN))/XM
	PE=1.D0-PD
	L=L+1
	ELM=DLOGIO(XM)
	X(L)=SNGL(ELM)
	ELM=DLOG10 (PE)
	Y(L)=SNGL(ELM)
	NPT(J)=NPT(J)+1

```
8Q/80 LIST
    04/30/68
   WRITE (NOUT,60) XM,PD,PE,X(L),Y(L)
60 FORMAT(9X,3(D16.9,6X),2(F16.9,6X))
65 CONTINUE
   65 CONTINUE
   70 CONTINUE
      L1=L+1
      L2=L+2
      X(L1)≈0.
      X(L2) = 0.
      Y(L1)=0.
      Y(L2)≈0.
CCC
      PLOTTING ROUTINES FOR CAL-COMP PLOTTER
      CALL PLOT(0.,-31.,-3)
      CALL PLOT(0.,2.,-3)
      CALL SCALE (X,XAX,L,1)
      CALL SCALE (Y, YAX, L, 1)
      CALL AXIS (0.,0.,6HLOG(M),-6,XAX,0.,X(L1),X(L2))
      CALL AXIS (0.,0,7HLOG(PE),7,YAX,90.,Y(L1),Y(L2))
      I1 = 0
      DO 90 I=1.NK
      NN=NPT(I)
      DO 85 J=1,NN
      I1 = I1 + 1
      EX(J) = X(II)
   85 WY(J)=Y(I1)
      NI = NN + I
      N2=NN+2
      EX(N1)=X(L1)
      EX(N2) = X(L2)
      WY(N1)=Y(L1)
      WY(N2)=Y(L2)
      CALL LINE(EX, WY, NN, 1,0,0)
   90 CONTINUE
      CALL SYMBOL(0.5,16.0,.3,3HKS=,0.,3)
      CALL NUMBER(1.5,16.0,.3,XKS,0.,-1)
      CALL PLOT(20.,0.,-3)
  110 CONTINUE
      IF(IEND.EQ.9) GO TO 1
      CALL FIN(IPLT)
  120 STOP
      END
```

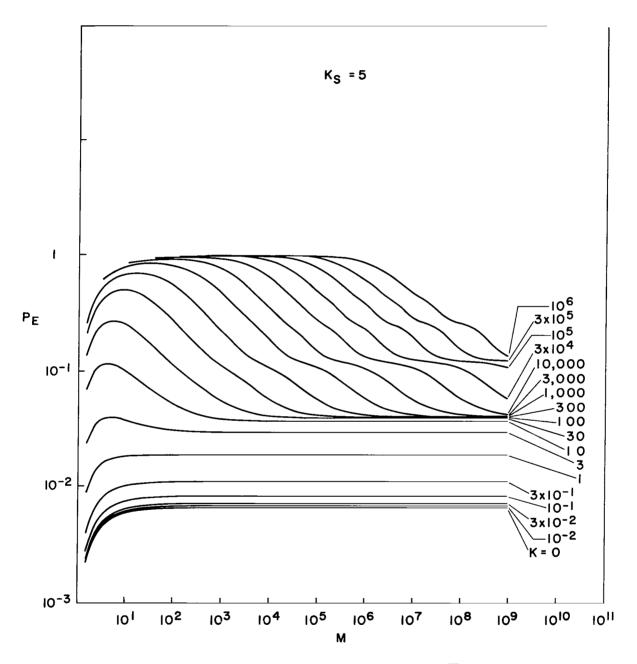


Figure 11.- Error probabilities for $\overline{n}_n T = K$ fixed as a function of M

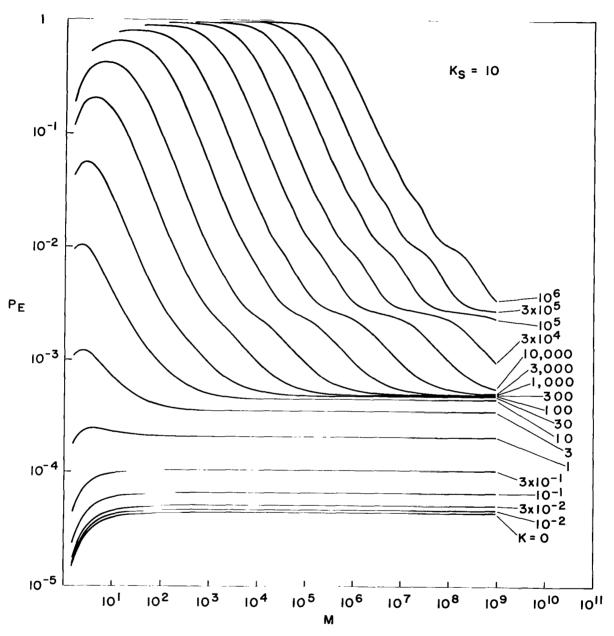


Figure 12.- Error probabilities for $\overline{n}_n T = K$ fixed as a function of M

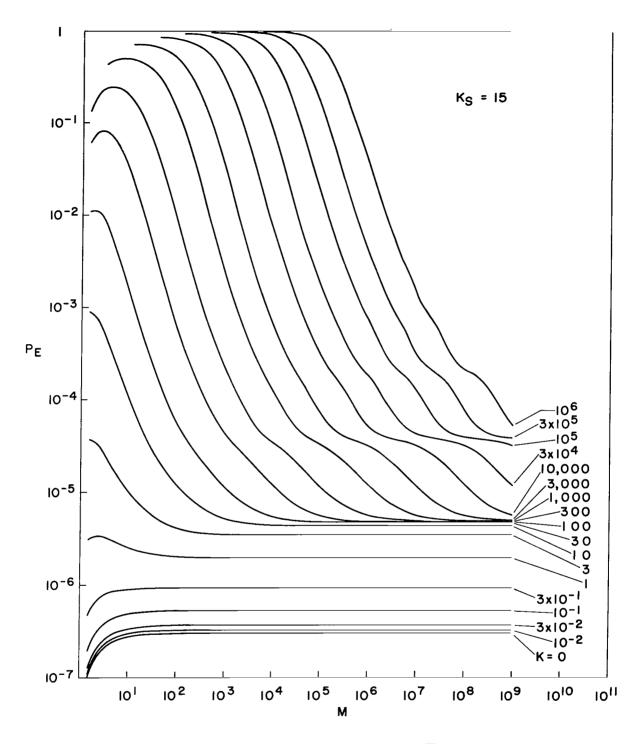


Figure 13.- Error probabilities for $\overline{n}_n T = K$ fixed as a function of M

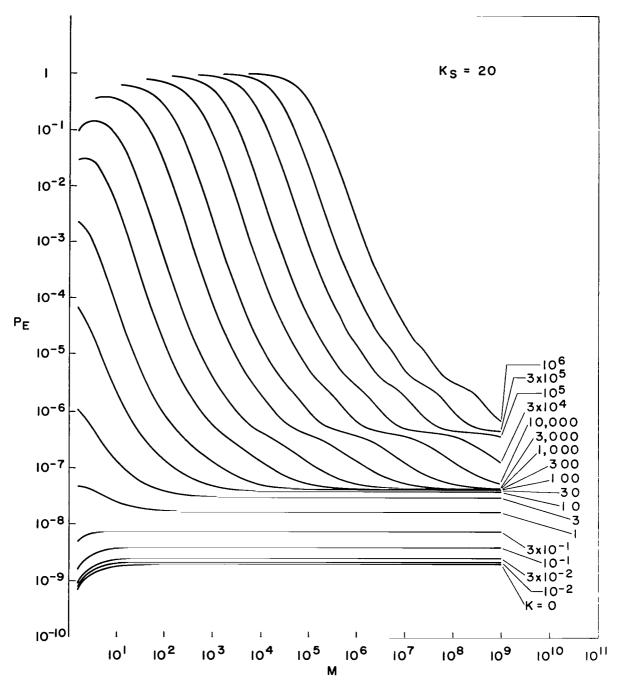


Figure 14.- Error probabilities for $\overline{n}_n T = K$ fixed as a function of M

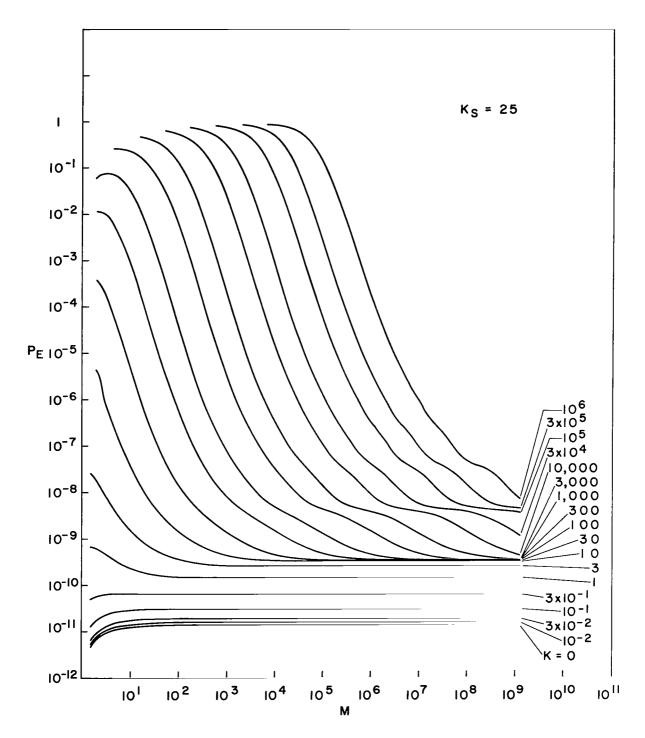


Figure 15.- Error probabilities for $\overline{n}_nT = K$ fixed as a function of M

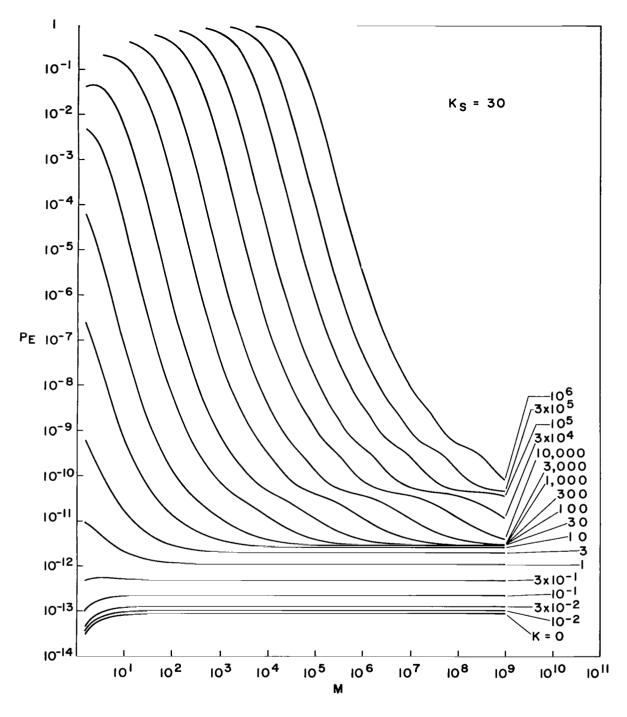


Figure 16.- Error probabilities for $\overline{n}_n T = K$ fixed as a function of M

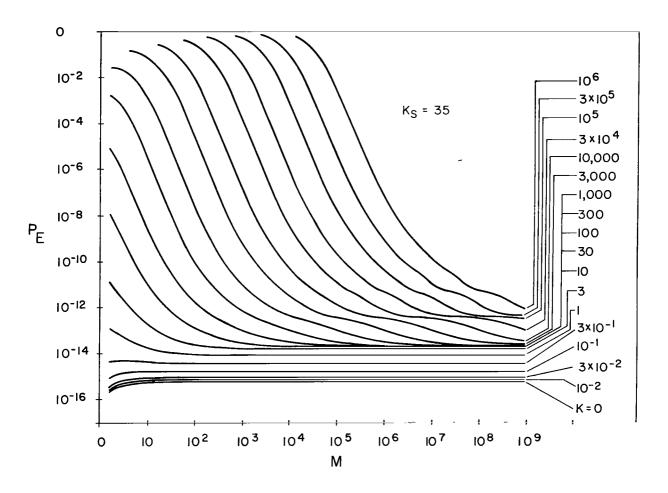


Figure 17.- Error probabilities for \overline{n}_nT = K fixed as a function of M

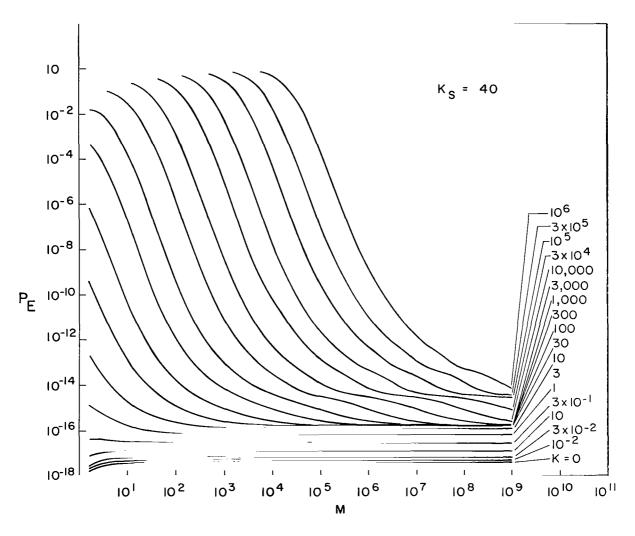


Figure 18.- Error probabilities for \overline{n}_nT = K fixed as a function of M

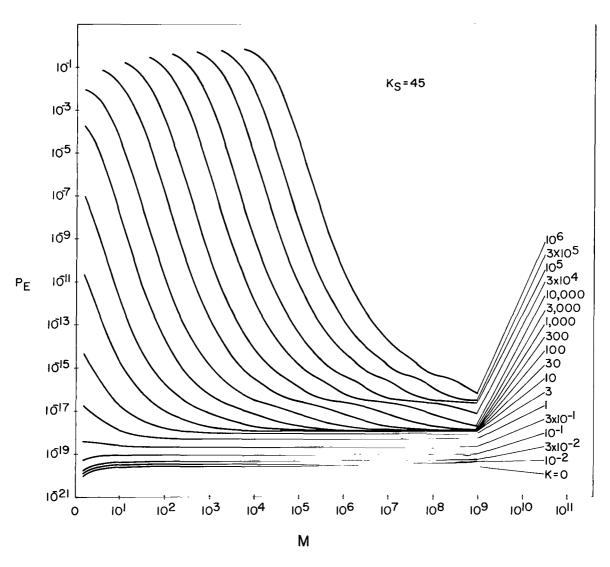


Figure 19.- Error probabilities for $\overline{n}_nT = K$ fixed as a function of M

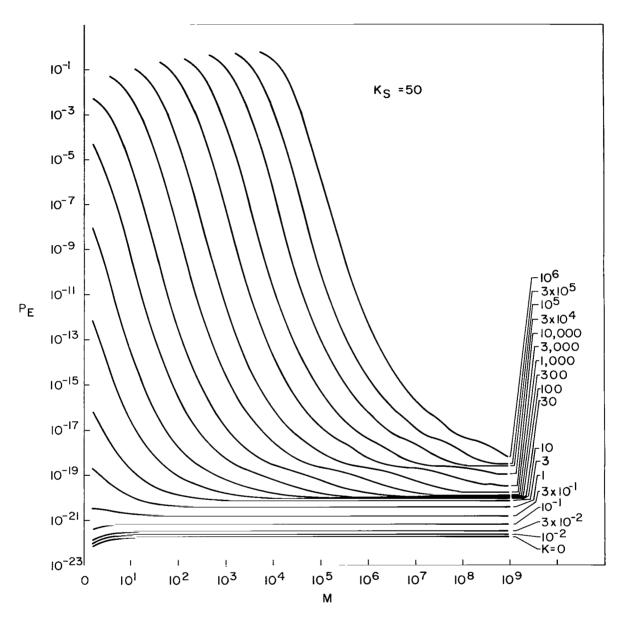


Figure 20.- Error probabilities for $\overline{n}_n T = K$ fixed as a function of M

COMPUTATIONS

The evaluation of $P_{\rm D}$ and $P_{\rm E}$ as functions of the parameters $K_{\rm s}$, $K_{\rm n},$ and M, was accomplished by a set of computer programs which are presented below.

In one case, K_s , K_n , and M are input to the program. In the second case K_n is computed as $K_n = K/M$ where K_s , K and M are input to the program.

The programs were written in Fortran IV and run on the CDC 6400 Computer System at the Smithsonian Astrophysical Observatory in Cambridge, Massachusetts.

DESCRIPTION OF THE FLOW CHART

The constants including tape definitions are set in data statements (Figures 21a,b, and c):

- Box 1: Input data and switch values. Set index J to 1. This index counts the number of K_n values.
- Box 2: Set index I to 1. This index counts the number of M values. Print headings for tables to be outputted.
- Box 3: Set the M value for this loop. In one version of the program K is fixed. In the other version it is computed as a function of M.
- Box 4: Test magnitude of K_n to prevent overflow or error.
- Box 5: Establish initial values of computed parameters.
- Box 6: Test magnitude of (1+B)^M to prevent overflow or error.
- Box 7: Compute value of term 3.
- Box 8: Test for K_n near zero.
- Box 9: Compute initial values for PI, Sum PI, Term 2.
- Box 10: Compute initial values for term, total.
- Box 11: Set index IX to 2. This counts number of times through major computation loop.
- Box 12: Compute PX.
- Box 13: Test for K_n near zero.
- Box 14: Compute PI, Sum PI, Term 2.
- Box 15: Compute new B.
- Box 16: Test magnitude of $(l + B)^{M}$.
- Box 17: Compute Term 3.

- Box 18: Compute new term and add this total.
- Box 19: Test for completion of major loop. If the difference between two successive values of term is sufficiently small, and the index has exceeded a predetermined cutoff value, the loop is completed.
- Box 20: Increment IX by 1.
- Box 21: Test for sufficient number of times through the loop.
- Box 22: P_D is set to final value of total. P_E is $1-P_D$.
- Box 23: Print out table of M, P_D , P_E and their logarithms.
- Box 24: Increment index I by I.
- Box 25: Test for maximum value of M.
- Box 26: Increment index J by 1.
- Box 27: If J has not exceeded J max, recycle for next value of K (or K_n).
- Box 28: Plot results.
- Box 29: Test whether any more data sets are to be computed.

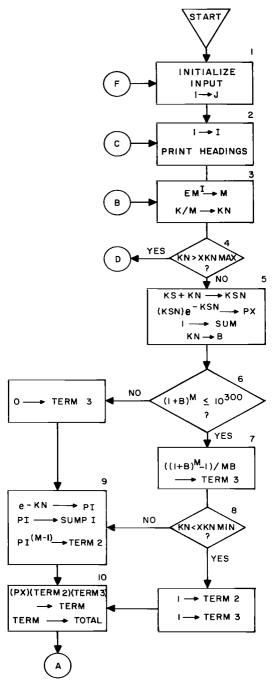


Figure 21a. - Program flow chart

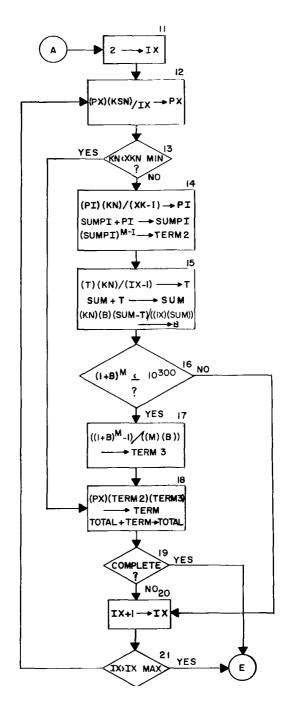


Figure 21b. - Program flow chart

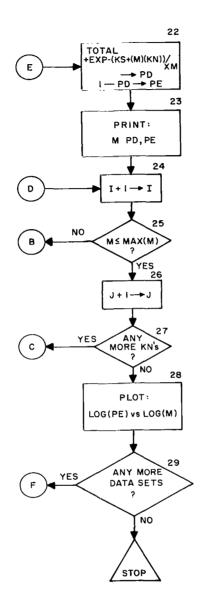


Figure 21c.- Program flow chart

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